



# Simple near-monostatic Mueller polarimeter

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# Simple near-monostatic Mueller polarimeter

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Inferred material properties from active or passive polarimetric remote sensing requires knowledge of the polarized bidirectional reflectance distribution function (P-BRDF). To encourage and simplify the process of measuring the P-BRDF of materials in the monostatic or near-monostatic configurations seen in lidar and other active remote sensing systems, we describe the design, build, and operating procedure for a simple near-monostatic polarimeter. The described system operates at 1064 nm and with a transmitter-receiver separation of about 4°, but the design principles, alignment procedure, and measurement procedure can be directly applied to systems at other wavelengths or separation angles. Also included are the control and data analysis software for the instrument we deployed, and a small selection of retrievals from the instrument.

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## 1. INTRODUCTION

Polarization represents a significant means of improving data density in combination with the more common retrieved light characteristics: intensity, spatial frequency, and spectral distributions. To date, polarization-enhanced remote sensing has focused primarily on passive methods [1] and dual-polarization lidar [2].

Polarization-enhanced lidar has been used for a wide variety of applications, especially in atmospheric and oceanic remote sensing [2]. It has also been used to improve the detection of submerged samples with sensors above the water [3]. Foliage and plant species discrimination are areas of interest in active polarimetric sample discrimination, including imaging polarimeters [4, 5], active lab-based polarimetry [6], and dual-polarization lidars [7]. There is also opportunity to improve material discrimination in complex scenes using polarimetry, as demonstrated in previous passive systems [8] and active systems [9, 10].

To more accurately identify samples in a scene, the community requires means to characterize reference materials specific to their applications, similar to those which exist for satellite or other passive remote sensing systems. Data from previous works report material polarimetric responses by their polarized bidirectional reflectance distribution function (P-BRDF) [6, 11–14]. However, of these reported P-BRDFs, only those in [13] in-

cluded measurements of the monostatic configuration, the most common configuration for lidar and similar active polarimetry. These monostatic P-BRDF (P-mBRDF) returns are important for polarization-enhanced lidar with applications moving toward material discrimination.

In this paper, we describe the design, operating procedure, and data analysis methods for a simple near-monostatic polarimeter. We also show some data collected by the instrument we built, the LabView collection software we developed, and the MATLAB data analysis script used to process the raw intensity data into Mueller matrix P-mBRDF plots.

## 2. BACKGROUND

For material discrimination, we are primarily interested in determining Mueller matrix,  $\mathbf{M}$ , values, which represent material properties. In active polarimetry, we use a polarization state generator (PSG) in combination with a light source, such as a laser, to illuminate the scene with known polarization characteristics,  $\mathbf{S}_G$ . Optical detectors are only directly capable of detecting the power or photon flux incident on their active surface, so a polarization state analyzer (PSA) with a known polarization response,  $\mathbf{M}_A$ , is used to attenuate total power as a function of the incident light's polarization state. Extended to an equation, the PSG-PSA active polarimetric system is described as

$$\mathbf{S}_{det} = \begin{bmatrix} S_{0,det} \\ S_{1,det} \\ S_{2,det} \\ S_{3,det} \end{bmatrix} = \begin{bmatrix} I_{det} \\ Q_{det} \\ U_{det} \\ V_{det} \end{bmatrix} = \mathbf{M}_A \mathbf{S}_{Rx} = \mathbf{M}_A (\mathbf{M}_{samp} \mathbf{S}_G), \quad (1)$$

where  $\mathbf{S}_{det}$  describes the light at the detector and  $\mathbf{M}_{samp}$  describes the polarization sensitive properties of the sample in the scene. As seen above, a Stokes vector can be broken down into its four components. The  $S_0$  or  $I$  element represents the total power regardless of polarization, and is the only element measured by the detector's active area. As such, we require a minimum of 16 unique measurements to fully reconstruct an unknown  $\mathbf{M}_{samp}$ . This is most often accomplished by setting four PSG states and four PSA states and performing measurements for each of the unique PSG-PSA state combinations.

Optimization of polarimetry systems primarily concerns proper selection of these PSG and PSA states. As shown in previous literature, optimal PSG states are those matched to the

PSA states [15, 16]. Selecting optimal PSA states is best conceptualized using the Poincaré sphere, where optimal polarization states have Poincaré coordinates which define a regular polyhedron [17, 18]; for four PSA states, this generates a regular tetrahedron.

Following the processes in previous literature, we can quantify the maximum error propagation for transmitter and receiver errors, including noise, by determining the condition number,  $\kappa$ , of the PSG and PSA system matrices [16, 18]. The condition number for the PSG,  $\kappa_G$ , represents the maximum error propagation of intensity or polarization noise in the Tx source. The condition number for the PSA,  $\kappa_A$ , represents the maximum error propagation of noise in the Rx detector.

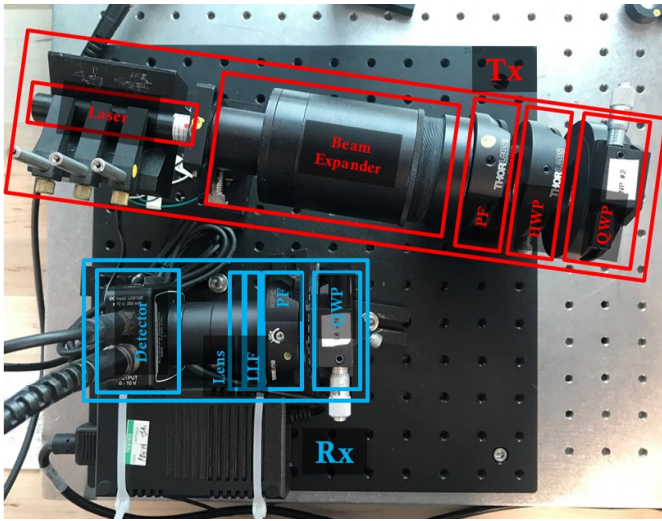
To find  $\kappa$  for each branch of the radiometer, we use

$$\kappa(\mathbf{A}) = \|\mathbf{A}\|_2 \|\mathbf{A}^{-1}\|_2 = \frac{\sigma_{\max}(\mathbf{A})}{\sigma_{\min}(\mathbf{A})}, \quad (2)$$

where  $\|\cdot\|_2$  is the  $L^2$  norm,  $\sigma_{\max}$  and  $\sigma_{\min}$  are the maximum and minimum singular value decompositions, and  $\mathbf{A}$  is the analysis matrix of the system's polarization elements, defined as

$$\mathbf{A} = \begin{bmatrix} \mathbf{M}_0^1 \\ \mathbf{M}_0^2 \\ \mathbf{M}_0^3 \\ \mathbf{M}_0^4 \end{bmatrix} = \begin{bmatrix} m_{00}^1 & m_{01}^1 & m_{02}^1 & m_{03}^1 \\ m_{00}^2 & m_{01}^2 & m_{02}^2 & m_{03}^2 \\ m_{00}^3 & m_{01}^3 & m_{02}^3 & m_{03}^3 \\ m_{00}^4 & m_{01}^4 & m_{02}^4 & m_{03}^4 \end{bmatrix}, \quad (3)$$

where  $\mathbf{M}_0^i$  is the first row of the combined Mueller matrix of the polarization elements for the PSG's or PSA's  $i$ th state. When  $A$  is singular (or near-singular) you may need to use the pseudo-inverse of  $A$  in Eq. 2 instead of the simple  $A^{-1}$  inverse. For an optimal configuration, the condition number of the PSG and PSA are each equal to  $\sqrt{M-1}$ , where  $M$  is the dimensionality of the polarization (3 for linear systems and 4 for full-polarimetric systems). Therefore, an optimal system will have  $\kappa_G = \kappa_A = \sqrt{2}$  for a linear polarimetry system, or  $=\sqrt{3}$  for a full-polarimetric system.



**Fig. 1.** Overhead image of the simple polarimeter, with the transmitter optics labelled in red and the receiver optics labelled in blue. This figure is a partial reproduction of a figure in [19].

### 3. INSTRUMENT DESIGN

#### A. Example Design

In this section, we walk through the design of the active polarimeter we built, and provide reasoning such that others could design similar systems at different wavelengths using the same principles. An overhead view of the system we built is shown in Fig. 1 with components labelled.

The instrument begins with a simple active radiometer: a laser transmitter (Tx) and a photodiode receiver (Rx). The system we built used a 10-mW, 1064-nm, diode-pumped solid-state laser and a ThorLabs PDA10CS2 photodiode, which is an In-GaAs, variable-gain detector with a 1-mm-diameter circular active area. To reduce speckle effects, the 1-mm-diameter laser beam was expanded by a 10x beam expander; this also increased the beam divergence from about 0.5 mrad to 5 mrad. For the receiver optics, we used a single 50-mm-focal-length plano-convex lens with a 25-mm clear aperture, resulting in a full field-of-view (FOV),  $\omega_{FOV}$ , of 10 mrad, which is about double the transmitter beam divergence. Placing the planar side toward the detector and the convex side toward the scene minimizes aberrations for the simple lens system. Finally, a 10-nm bandwidth laser-line spectral filter (LLF) for a 1064-nm central wavelength was placed on the scene side of the receiving optics to eliminate the majority of background light from the scene.

The PSG was inserted in front of the transmitter and consisted of a linear polarizing filter (PF), a half-wave retarder (HWP), and a quarter-wave retarder (QWP). The PF was locked to transmit vertical linear polarization, and the laser was rotated in its mount until maximum transmittance through the filter was achieved. This provided a much purer initial polarization state than the laser alone, since the PF used has an extinction ratio of approximately  $10^3$ . The HWP was set in a rotation mount, which allowed us to rotate the initial linear state to other orientation angles without attenuation. The QWP was set with its fast axis at  $45^\circ$  on a  $90^\circ$  flip-mount, allowing it to be inserted into or removed from the optical path quickly and easily.

The PSA was inserted in front of the receiver optics and consisted only of a QWP and a PF. The QWP was set to  $-45^\circ$  on a  $90^\circ$  flip-mount for the same purpose as in the PSG. The PF was set in a rotation mount to allow for preferential attenuation for different linear orientation angles.

For an analog-digital converter (ADC), we used a National Instruments USB-6001 data acquisition device. For a total list of parts, excluding some minor optomechanical mounting equipment, see Table 1.

#### B. System Analysis

For the purposes of compiling and potentially sharing reference measurements, it's important to know the specifications of the systems acquiring the reference data. Some critical system information includes the operating wavelength, actual geometric separation angle between transmitter and receiver, receiver FOV, and other specifications for which we give an example summary in Table 2.

The operating wavelength is critical information as many materials exhibit reflectance and transmittance characteristics which are strongly spectrally dependent. To a lesser extent, the receiver optical bandwidth can also be useful information when considering the spectral effects on the sample characteristics. The transmitter beam diameter at the output, beam divergence, and beam diameter at the sample help inform the probability of speckle error and the total illumination area considered for

**Table 1. Developed System Parts and Approximate Costs**

Part	Cost, \$
Laser (Edmund Optics #37-043)	350
Detector (ThorLabs PDA10CS2)	475
PF (Meadowlark DPM-100-NIR2-1) × 2	2 × 685
HWP (Meadowlark NHM-100-1064)	575
QWP (Meadowlark NQM-100-1064) × 2	2 × 575
ADC (NI USB-6001)	300
Rotation Mount (ThorLabs RSP1) × 5	5 × 100
Flip Mount (ThorLabs FM90) × 2	2 × 100
Lenses	100
total	5,020

152 samples with spatial anisotropy. The viewing solid angle,  $\Omega_v$ ,  
 153 which should be relatively narrow for the purpose of P-mBRDF  
 154 characterization, is determined by the smaller of the solid angles  
 155 produced by Eq. 4 or Eq. 5.

$$\Omega_v^{FOV} = 4\pi \sin^2\left(\frac{\omega_{FOV}}{4}\right) \quad (4)$$

$$\Omega_v^A = \frac{A_s}{Z^2} = \frac{\pi D_s^2}{4Z^2} \quad (5)$$

157 Finally, the actual geometric separation angle between the  
 158 transmitter and receiver is critical to report the actual monostatic  
 159 or near-monostatic nature of the characterizing measurement.

160 The signal-noise ratio (SNR) for the developed system was  
 161 calculated using a theoretical signal for a 10% Lambertian reflector  
 162 against the detector noise equivalent power (NEP). In a lab  
 163 setting, the optical power incident on the detector,  $P_{Rx}$ , follows a  
 164 basic radiometric equation,

$$P_{Rx} = A_{Rx} \Omega_v L_s = \frac{\pi D_{Rx}^2}{4} \Omega_v \frac{4I_{Tx} \rho_s T_{atm}}{\pi^2 D_s^2} = \frac{D_{Rx}^2 \Omega_v I_{Tx} \rho_s T_{atm}^2}{\pi D_s^2}, \quad (6)$$

165 where  $\rho_s$  is the reflectance of a Lambertian sample reflector and  
 166  $T_{atm}$  is the transmittance of the atmosphere (typically about 1 in  
 167 a lab setting). For the developed system,  $P_{Rx}$  comes out to about  
 168  $0.269 \mu W$ , and the maximum NEP for the detector is  $0.166 \mu W$ ,  
 169 giving an initial SNR of 1.62, or equivalently, 4.2 dB. The ADC  
 170 has a sampling rate of 20 kS/s, so averaging over 0.5 seconds  
 171 gives 10,000 samples, and improves the SNR by a factor of 100,  
 172 to about 162, or equivalently, 44.2 dB.

173 For ease of use, the nearest-optimum states we can select are  
 174 a circular state (RHC) and three linear states with orientations  
 175 equally separated by  $60^\circ$ , which we select as  $0^\circ$ ,  $60^\circ$ , and  $120^\circ$ .  
 176 Using these states for both PSG and PSA, the condition numbers  
 177 are  $\kappa_G = \kappa_A \approx 2.48$ . This means any output intensity noise after  
 178 the clean-up polarizer, whether due to intensity or polarization  
 179 instability in the laser, can propagate to the polarization mea-  
 180 surements with a maximum magnification of 2.48. Similarly,  
 181 noise in the detector can propagate to Mueller matrix element  
 182 calculation with a maximum magnification of 2.48. These  $\kappa$   
 183 values are only about 43% greater than those for the optimal  
 184 Mueller polarimeter,  $\sqrt{3} \approx 1.73$ .

185 The maximum error propagation reduces the SNR of the  
 186 polarimetric results as compared to the system SNR.

$$SNR_{pol} = \frac{SNR_{sys}}{\kappa_A} \quad (7)$$

The values are summarized in Table 2.

**Table 2. Developed System Specifications**

Parameter	Value
Operating wavelength, $\lambda$	1064 nm
Receiver optical bandwidth, $\Delta\lambda$	$\pm 5$ nm
Transmitter optical power, $I_{Tx}$	10 mW
Beam diameter at output, $D_b$	10 mm
Beam divergence, $\omega_{div}$	5 mrad
Beam diameter at sample, $D_s$	13.8 mm
Distance to sample, $Z$	762 mm
Receiver aperture diameter, $D_{Rx}$	25 mm
Full FOV, $\omega_{FOV}$	10 mrad
Tx-Rx separation angle, $\theta_\Delta$	$4^\circ$
Sample rotation axis	y-axis (vertical)
System signal-noise ratio, $SNR_{sys}$	162 (44.2 dB)
Max error propagation, $\kappa_G = \kappa_A$	2.48
Min polarimetric SNR, $SNR_{pol}$	65.3 (36.3 dB)

## 4. OPERATING PROCEDURE AND DATA ANALYSIS

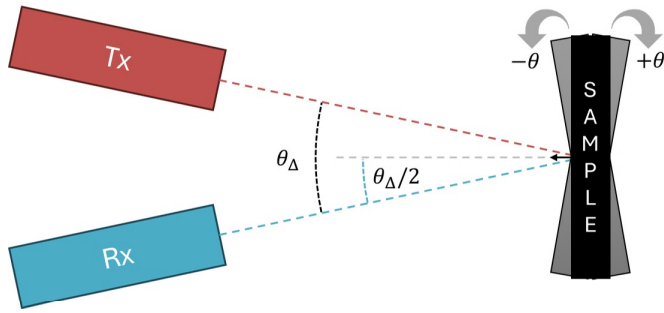
### A. Optomechanical Alignment

187 Alignment of the polarimeter optical trains is relatively straight-  
 188 forward. Use multiple irises to align a beam expander or similar  
 189 optics along the transmitter optical path. For the receiver, a lens  
 190 tube attached to the detector containing the focusing lens at its  
 191 focal length from the detector active area is typically sufficient  
 192 for axial alignment. Measuring the focal length separation to  
 193 within about 1% error is generally sufficient for this type of  
 194 radiometry as it is non-imaging.

195 After inserting an LLF, or while in a dark lab, but before  
 196 inserting the polarization elements, use a diffuse reflectance  
 197 sample and rotate either the transmitter or receiver optical paths  
 198 until you receive a maximum signal. Lock the transmitter and  
 199 receiver in place and note the geometric separation angle,  $\theta_\Delta$ , of  
 200 their optical paths. Also, note the maximum signal you achieved;  
 201 this will help in polarizer alignment.

202 The angular geometry used in the data we gathered is sum-  
 203 marized in Fig. 2, which is useful for similar fixed-separation  
 204 systems. The reference  $0^\circ$  angle is where the sample surface  
 205 normal is pointed halfway between the optical axes of the trans-  
 206 mitter and receiver. Rotations of the surface normal toward the  
 207 transmitter are referred to with positive angles, and rotations  
 208 toward the receiver are referred to with negative angles.

209 Next, insert the transmitter PF with its transmission axis as  
 210 horizontal as possible; most rotation mounts should allow for  $1^\circ$   
 211 ( $17.5$  mrad) tolerance or better on this orientation angle. For the  
 212 purposes of this procedure, horizontal orientation is  $0^\circ$ , vertical  
 213  
 214  
 215



**Fig. 2.** Overhead diagram depicting the geometric separation angle, the surface-normal  $0^\circ$  reference setup, and the general orientation of positive and negative viewing/illumination angles.

orientation is  $90^\circ$ , and  $45^\circ$  orientation is above horizontal and to the right of vertical when looking into the beam. If possible, rotate the laser head to maximize the power through the polarizer, as measured by the aligned receiver. Otherwise, you might consider rotating the PF to achieve max power throughput, shifting the suggested orientation angles of each of the other polarizer elements by the same amount. Lock the transmitter PF in place.

Then insert the receiver PF and rotate it until you have a minimum signal at the detector. Note the rotation angle as  $90^\circ$ , or otherwise at  $90^\circ$  from the transmitter PF orientation.

Leaving the receiver PF at this crossed orientation, insert the transmitter HWP and rotate it until you return to a minimum signal at the detector. Note this HWP orientation as the crystal axis aligned to the transmitter PF orientation ( $0^\circ$  by default).

The transmitter QWP should be labelled with a fast-axis (sometimes referred to as the crystal axis). Insert the transmitter QWP and rotate it so that the fast axis is at about  $-45^\circ$  from the transmitter PF orientation. Alternate the receiver PF between  $0^\circ$  and  $90^\circ$  (co- and cross-polarized with the transmitter PF) while rotating the QWP in small increments about  $-45^\circ$  until the co- and cross-polarized signals are equal. This produces right-hand circular (RHC) polarization. Lock the transmitter QWP in place, then flip it out of the optical path and repeat this process with the receiver QWP but at and about  $+45^\circ$  orientation.

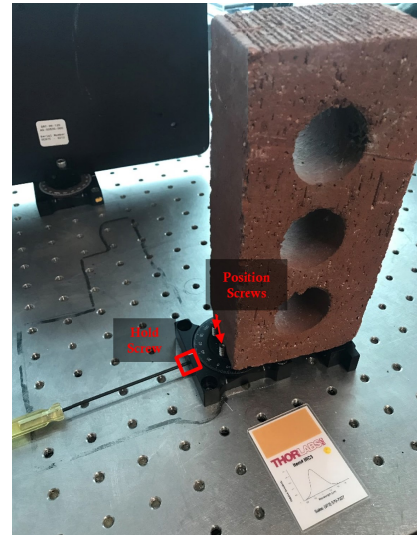
## B. Control & Analysis Software

The control software is programmed in LabView 2019. The software relies on drivers for the Newport USB-6001, but otherwise functions using basic LabView controls. This software is available in Code 1 (Ref. [20]).

The analysis software is programmed in MATLAB 2023a. This software loads .csv files as saved by the control software, reports the Mueller matrices as figures, and records the Mueller matrices as a function of angle in new .csv files, with each row of the file in the form  $[\theta_v m_{00} m_{01} m_{02} m_{03} m_{10} \dots m_{32} m_{33}]$ . This software is available in Code 2 (Ref. [21]).

## C. Data Collection Procedure

To characterize the P-mBRDF (or a fixed Tx-Rx separation P-BRDF) of a sample, either the sample or polarimeter must be rotated with respect to the other. In many cases, a fixed polarimeter with a rotating sample provides the most simply realized



**Fig. 3.** Image of a red clay brick sample on a horizontal rotation stage. Two set screws make it easy to return the brick to the same position repeatably, and to more accurately determine the incidence/viewing angle,  $\theta_v$ , relative to the polarimeter. This figure is a partial reproduction of a figure in [19].

setup. We mounted samples on a stage rotating about the y-axis (vertical axis). The samples were set at about 0.75 m from the front of the transmitter and receiver (see Table 2 for specific values).

When necessary, samples are prepared so they are planar. For example, a Spectralon reference or brick are rigid materials with a tall, flat face which does not need to be prepared. However, a roofing shingle or plant leaf are not rigid, and so must be attached to a backing surface such as a piece of wood in preparation. Once on a rigid surface of sufficient height, a horizontal rotation stage can be used as shown in Fig. 3 to make repeatable sample rotation much easier.

Once the sample is prepared and in place, follow the semi-time-optimized procedure here in combination with the control software noted above:

1. Set the sample at a known incidence angle.
2. Set the  $HWP_G$  to  $0^\circ$  and the  $PF_A$  to  $0^\circ$ . Flip the  $QWP_G$  and  $QWP_A$  out of the optical path.
3. In the control software, SCAN for Stage 1.
4. Flip the  $QWP_A$  into the beam path. SCAN for Stage 2.
5. Flip the  $QWP_G$  into the beam path. SCAN for Stage 3.
6. Flip the  $QWP_A$  out of the beam path. SCAN for Stage 4.
7. Rotate the  $PF_A$  to  $60^\circ$ . SCAN for Stage 5.
8. Flip the  $QWP_G$  out of the beam path. SCAN for Stage 6.
9. Rotate the  $PF_A$  to  $-60^\circ$ . SCAN for Stage 7.
10. Flip the  $QWP_G$  into the beam path. SCAN for Stage 8.
11. Flip the  $QWP_G$  out of the beam path, rotate the  $HWP_G$  to  $-30^\circ$ . SCAN for Stage 9.
12. Rotate the  $PF_A$  to  $60^\circ$ . SCAN for Stage 10.
13. Rotate the  $PF_A$  to  $0^\circ$ . SCAN for Stage 11.
14. Flip the  $QWP_A$  into the optical path. SCAN for Stage 12.
15. Rotate the  $HWP_G$  to  $30^\circ$ . SCAN for Stage 13.

- 288 16. Flip the QWP<sub>A</sub> out of the optical path. SCAN for Stage 14.  
 289 17. Rotate the PF<sub>A</sub> to 60°. SCAN for Stage 15.  
 290 18. Rotate the PF<sub>A</sub> to -60°. SCAN for Stage 16.  
 291 19. Set the sample to a new known incidence angle and repeat  
 292 steps 2-19 until you have completed your angular analysis.

293 In this procedure, the polarizing elements are distinguished as  
 294 being part of the PSA or PSG with a subscript *A* or *G*, respec-  
 295 tively. For example, QWP<sub>A</sub> is the QWP in the PSA (receiver)  
 296 arm of the polarimeter.

297 For our data collection efforts, we typically scanned from -60°  
 298 to 60°, or a smaller range on one side or the other for geometries  
 299 where symmetry is expected or previously demonstrated. These  
 300 angular limitations are primarily due to the illumination area  
 301 overflowing some of the smaller samples as their projected area  
 302 decreases.

303 Using this procedure, we have found that an experienced  
 304 operator can complete a 16-measurement set in about 1 to 3 min-  
 305 utes, and a new operator can typically complete a measurement  
 306 set in about 6-8 minutes. The total measurement time for a sam-  
 307 ple then depends on how many viewing/illumination angles  
 308 are considered.

#### 309 D. Data Structure and Analysis

310 The control software records measurement data as .csv files  
 311 containing  $N \times 17$  matrices, where  $N$  is the number of angles  
 312 for which measurement sets were recorded. Each row is a  
 313 17-element matrix, with the first element giving the incidence-  
 314 viewing angle,  $\theta_v$ , and the remaining 16 elements giving the  
 315 recorded signal (in mV) for each of the unique PSG-PSA states.  
 316 The states are given in the order of vertical-vertical, vertical-60°,  
 317 vertical-120°, vertical-circular, 60°-vertical, ... circular-120°, and  
 318 circular-circular. This is the structure expected by the analysis  
 319 software.

320 The data analysis is performed using linear regression to  
 321 reduce processing time. The fundamental principle follows from  
 322 a simple Mueller calculus equation,

$$323 \vec{S}_{g,a} = \begin{bmatrix} I_{g,a} \\ Q_{g,a} \\ U_{g,a} \\ V_{g,a} \end{bmatrix} = \mathbf{M}_a \mathbf{M}_{\text{samp}} \vec{S}_g, \quad (8)$$

324 where  $\vec{S}_{g,a}$  is the Stokes vector of light reaching the detector,  $\mathbf{M}_a$   
 325 is the analysis Mueller matrix,  $\mathbf{M}_{\text{samp}}$  is the unknown Mueller  
 326 matrix response of the sample, and  $\vec{S}_g$  is the Stokes vector of the  
 327 light generated in the transmitter arm. Since the detector can  
 only measure intensity, the equation extends out to

$$328 I_{g,a} = I_g(m_{a00}m_{s00} + m_{a01}m_{s10} + m_{a02}m_{s20} + m_{a03}m_{s30}) \\
+ Q_g(m_{a00}m_{s01} + m_{a01}m_{s11} + m_{a02}m_{s21} + m_{a03}m_{s31}) \\
+ U_g(m_{a00}m_{s02} + m_{a01}m_{s12} + m_{a02}m_{s22} + m_{a03}m_{s32}) \\
+ V_g(m_{a00}m_{s03} + m_{a01}m_{s13} + m_{a02}m_{s23} + m_{a03}m_{s33}), \quad (9)$$

a simple linear combination equivalently represented as

$$329 I_{g,a} = B_{g,a} \vec{A}, \quad (10)$$

where

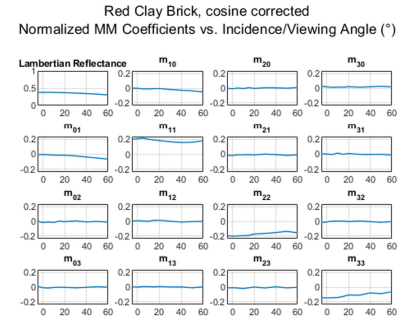
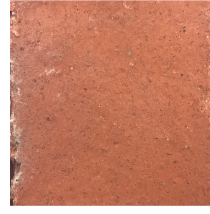


Fig. 4. An image of the sample brick surface about the mea-  
 surement location (left) and the resulting P-mBRDF measure-  
 ment (right).

$$\vec{A} = \text{vec}(\mathbf{M}_{\text{samp}}) = [m_{s00} \ m_{s10} \ m_{s20} \ m_{s30} \ m_{s01} \ m_{s11} \ m_{s21} \ m_{s31} \\
m_{s02} \ m_{s12} \ m_{s22} \ m_{s32} \ m_{s03} \ m_{s13} \ m_{s23} \ m_{s33}]^T \quad (11)$$

330 and

$$331 \vec{B}_{g,a} = [I_g m_{a00} \ I_g m_{a01} \ I_g m_{a02} \ I_g m_{a03} \\
332 Q_g m_{a00} \ Q_g m_{a01} \ Q_g m_{a02} \ Q_g m_{a03} \\
333 U_g m_{a00} \ U_g m_{a01} \ U_g m_{a02} \ U_g m_{a03} \\
334 V_g m_{a00} \ V_g m_{a01} \ V_g m_{a02} \ V_g m_{a03}]. \quad (12)$$

335 Extending Eq. 10 to the 16 measurements for each unique com-  
 bination of  $g$  and  $a$ , we can expand the scalar  $I_{g,a}$  to the  $1 \times 16$   
 vector  $\vec{I}_{\text{meas}}$ , and expand the  $1 \times 16$  vector  $\vec{B}_{g,a}$  to the  $16 \times 16$  matrix  
 $\mathbf{B}$ . Then, solving for  $\vec{A}$ , which is a simple vectorization of the  
 sample Mueller matrix, we find

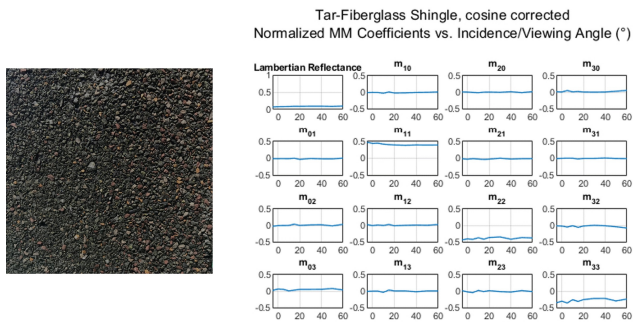
$$\vec{A} = \mathbf{B}^{-1} \vec{I}_{\text{meas}}. \quad (13)$$

## 336 5. EXAMPLE MATERIAL MEASUREMENTS

337 Some example data recorded with the example system design  
 338 are presented here. These data represent the near-monostatic  
 339 response of a red clay brick sample and a tar-fiberglass shingle  
 340 sample, as previously reported in [19]. These data also have  
 341 their  $m_{00}$  elements normalized to the near-normal returns of a  
 342 Spectralon Lambertian reference of known reflectance. The plots  
 343 shown here are also available in greater resolution in [19].

344 The red clay brick sample, shown in Fig. 4, was relatively  
 345 matte and of middling reflectance at the operating wavelength.  
 346 As such, we expect relatively flat reflectance across  $\theta_v$ , with a  
 347 slight dip at steeper angles and a moderate increase in polar-  
 348 ization and diattenuation at steep angles as well. These features  
 349 can be observed in the results shown in Fig. 4, and we can also  
 350 see that the off-diagonal elements for rotation and phase offset  
 351 are minimal relative to the diagonal elements.

352 A tar-fiberglass shingle was also measured, as seen in Fig. 5.  
 353 This sample was the least in-line with expected Lambertian or  
 354 specular surface responses. The shingle consisted primarily of  
 355 1- to 3-mm diameter rough fragments of tar overlaying a thin  
 356 layer of fiberglass insulation which was still visible in places  
 357 through the tar. The reflectance was fairly Lambertian, at least in  
 358 line with the brick's. Likely due to the high anisotropy of the tar



**Fig. 5.** An image of the sample tar-fiberglass shingle surface about the measurement location (left) and the resulting P-mBRDF measurement (right).

fragments, there was also no significant increase in polarizance and diattenuation at steeper  $\theta_v$ . For a rough surface, we would expect a reduction in polarization maintenance (the diagonal elements), but we see relatively strong maintenance in this sample. Dark surfaces and specular reflectors both have stronger polarization maintenance, and it is unclear for such a complex sample if this effect is due to the darkness of the tar or the high reflectance of the visible regions of fiberglass.

## 6. DISCUSSION

We have discussed the purpose, theory, and design principles for simple monostatic and near-monostatic polarimeters. The example design and cited software serve to enable groups with shared interest in active polarimetry for sample discrimination to develop their own reference measurements.

One of the major limitations of our example system is that it records only at 1064-nm wavelength. As spectral characteristics of materials change, their polarimetric responses are all but certain to change as well. For groups aiming to make a simple instrument to characterize materials for other devices at specific operating wavelengths, such as a lidar operating at a single wavelength, this narrow spectral reference will work well so long as the simple polarimeter has the same wavelength as the main device.

The other limitation of our example system is that it is not truly monostatic. The system for which we were generating sample references was not truly monostatic, so a near monostatic was sufficient for our research. If, however, a true monostatic geometry is preferred, a group could use a non-polarizing beam-splitter in front of the Tx output which reflects some of the Tx signal into a beam-block, then reflects a fraction of the return signal out of the primary optical path to a Rx setup on the reflected path. This option produces a true-monostatic geometry, but may complicate design due to higher power requirements to produce similar SNR.

Another consideration, given the longer time scale of the measurement sets, is the transmitter power stability, and by extension, its polarization stability. This is relatively simply measured by placing the transmitter source behind the clean-up polarizer and firing into a detector, taking constant measurements for about 15 to 20 minutes. The laser need only be stable (not changing significantly) over the time period typical of a 16-measurement set; about 1-3 minutes for an experienced operator

or 6-8 minutes for a newer operator.

Though it results in a much more expensive system, motorized rotation stages can be used in place of the manual rotation stages. In such a system, it is similarly efficient to realize a polarimetrically optimized set of PSG-PSA state combinations. For such a setup, we recommend the following four states for the unique states of the PSG and PSA individually, reported as their Poincaré coordinates  $(2\theta_P F, \phi_\Delta, DoP)$ :  $(0^\circ, 90^\circ, 1)$ ,  $(0^\circ, -19.4712^\circ, 1)$ ,  $(120^\circ, -19.4712^\circ, 1)$ , and  $(-120^\circ, -19.4712^\circ, 1)$ .

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**Data Availability Statement.** Data underlying the results presented in this paper are available to view in Ref. [19]. Additional data underlying the results presented in this paper are not publicly available at this time due to continuing efforts to create a publicly accessible database of P-mBRDF references, but our data may be obtained from the authors upon reasonable request.

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